

# HA17524P/FP

## Switching Regulator Controller

# HITACHI

ADE-204-058 (Z)

Rev. 0

Dec. 2000

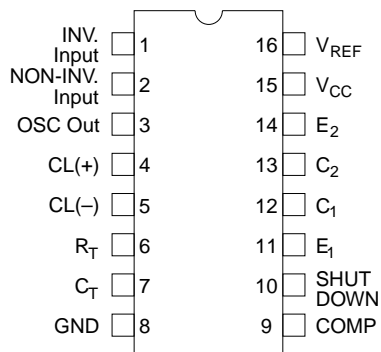
### Features

- Pulse width modulation (PWM)
- Wide oscillation frequency range: 450 kHz(typ)
- Low quiescent current: 5 mA typ
- Good line regulation (0.2% typ) and load regulation (0.4% typ)
- Independent output stages for 2 channels
- Wide external circuit applications including single-end and push-pull method
- Reference power source output stage and switching output stage include current limiting protection circuit.

### Ordering Information

Type No.	Package
HA17524P	16 pin dual in line plastic(DP-16)
HA17524FP	16 pin flat plastic (FP-16DA)

### Pin Arrangement



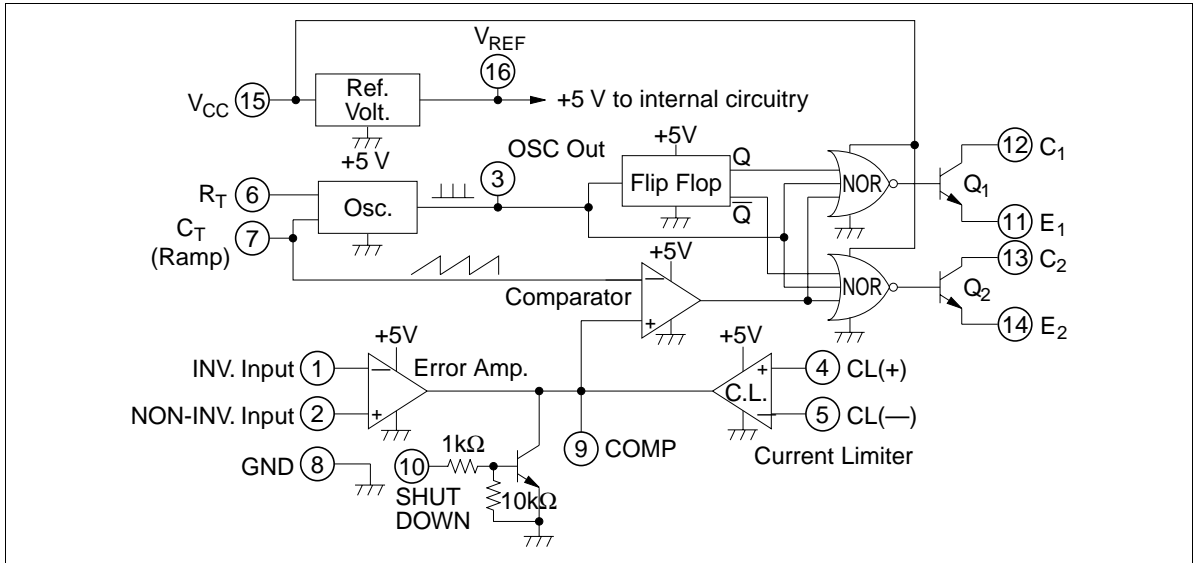
(Top View)

## Functional Description

### Principals of HA17524 Operation

The HA17524 switching regulator circuit, using pulse width modulation (PWM), is constructed as shown in figure 1.

Timing resistances  $R_T$  and timing capacitance  $C_T$  control the oscillation frequency.  $C_T$  is charged by a constant current generated by  $R_T$ . Ramp signals (saw-tooth waves) at the  $C_T$  terminal generated by this oscillator is available for reference input signal to comparator which control the pulse width.



**Figure 1 HA17524 Block Diagram**

The reference voltage connects to the non-inverted or inverted input terminal of the error amplifier via resistance divider (figure 2).

The output voltage from the error amplifier is compared with the ramp signal capacitance  $C_T$  (figure 1). The comparator can provide a signal with modulated pulse width.

This signal, then, controls output transistors  $Q_1$  and  $Q_2$ , making an open loop to stabilize output voltage.

Outputs from the error amplifier, the current limiter, and the shut-down circuit are connected together at the comparator, so that an input signal from any one of these circuits can break the output stage.

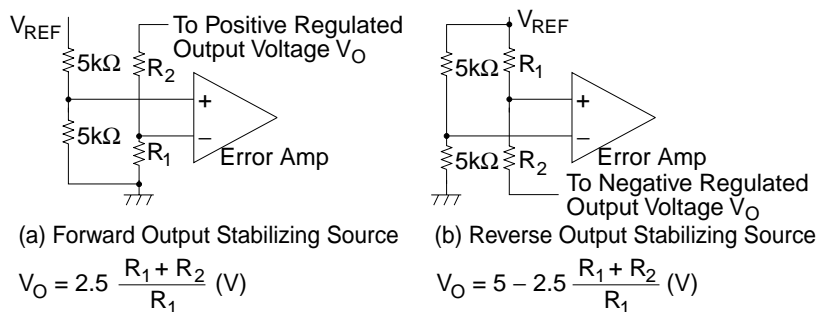


Figure 2 Error Amplifier Biasing

**Blocks Description**

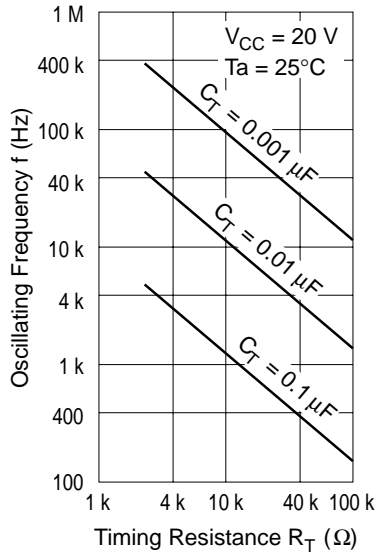
Oscillator: The oscillation frequency  $f$  is calculated from the following equations. Figure 3 shows one example.

$$f = 1.15 / (R_T \cdot C_T)$$

$$R_T = 1.8\text{k} \text{ to } 100\text{k} \Omega$$

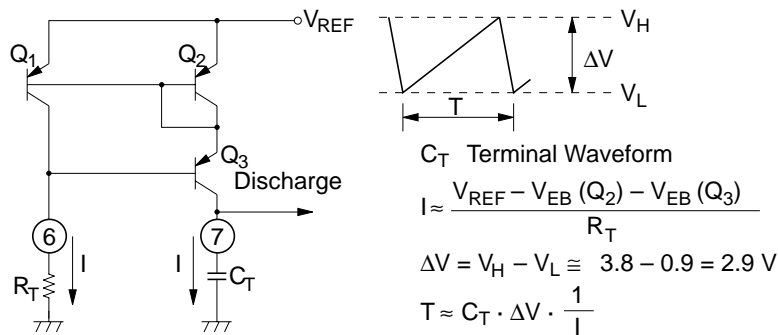
$$C_T = 0.001\mu \text{ to } 0.1 \mu\text{F}$$

$$f = 140\text{ Hz} \text{ to } 500\text{ kHz}$$



**Figure 3 Oscillating Frequency vs Timing Resistance**

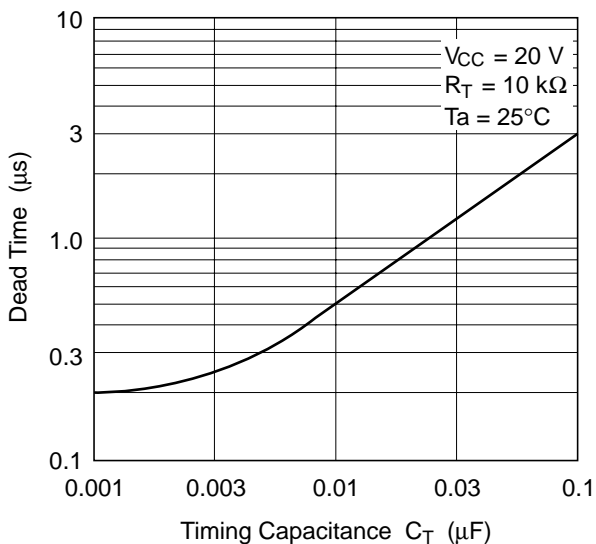
Then the ramp wave shown in figure 4 is available at pin 7,  $C_T$  terminal, since  $C_T$  is charged by the constant current  $I$  generated by  $R_T$ .



**Figure 4 Oscillating Circuit and  $C_T$  Terminal Waveform**

The oscillator output pulse signal is used as the flip flop clock pulse and as switching pulses for the output transistors, synchronous to the clock pulse.

The pulse-widths which can be controlled by the timing capacitor  $C_T$  as shown in figure 5, increases output dead time.



**Figure 5 Dead Time vs Timing Capacitance**

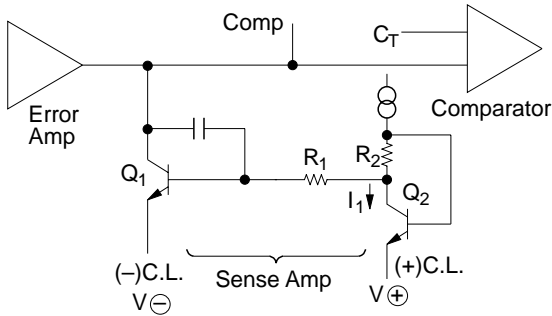
**Reference Voltage:** The built-in regulator (reference voltage:  $V_{REF} = 5 \pm 0.4 \text{ V}$ ) can be used as a reference power supply for the error amplifier, which determines output voltage ( $V_{OUT}$ ). It is also connected as a bias source for another circuits in IC.

**Error Amplifier:** Figure 2 shows error amplifier biasing, applied input voltage must be set within the range of common-mode input voltage (1.8 V to 3.4 V). Inserting a resistor and capacitor between phase compensation terminal (pin 9) and GND in series provides phase compensation.

**Current Limiter:** The sense amplifier threshold voltage ( $V_S$ ) for the current limiter is:

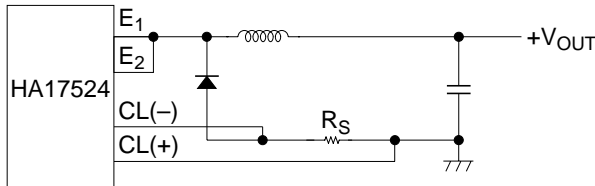
$$\begin{aligned} V_S &= V_{BE}(Q_1) + I_1 R_2 - V_{BE}(Q_2) \\ &= I_1 R_2 \\ &= 200 \text{ mV typ} \end{aligned}$$

At the current limiter sense amp shown in figure 6, when  $V^+ - V^- = 200 \text{ mV}$ ,  $Q_1$  turns on, phase compensation terminal becomes low and the output switching element is cut off.



**Figure 6 Current Limiter Sense Amplifier**

Figure 7 shows an example of detecting current limit. The input voltage range is  $-0.7 \text{ V}$  to  $+1.0 \text{ V}$ ; The current limit detection output is provided from GND line.



$$\begin{aligned} I_{OS} &= \frac{V_S}{R_S} \\ V_S &= 200 \text{ mV} \end{aligned}$$

**Figure 7 Current Limit Detector Example Operating Waveforms**

Operating Waveforms

Figure 9 shows operating waveforms at every part, when stepdown voltage type chopper switching regulator (figure 8) is used. Operating condition are as follows:  $f = 20 \text{ kHz}$ ,  $V_{\text{OUT}} = 5 \text{ V}$ . At the output section, two channels are connected in parallel. Operating waveforms inside the IC are also shown.

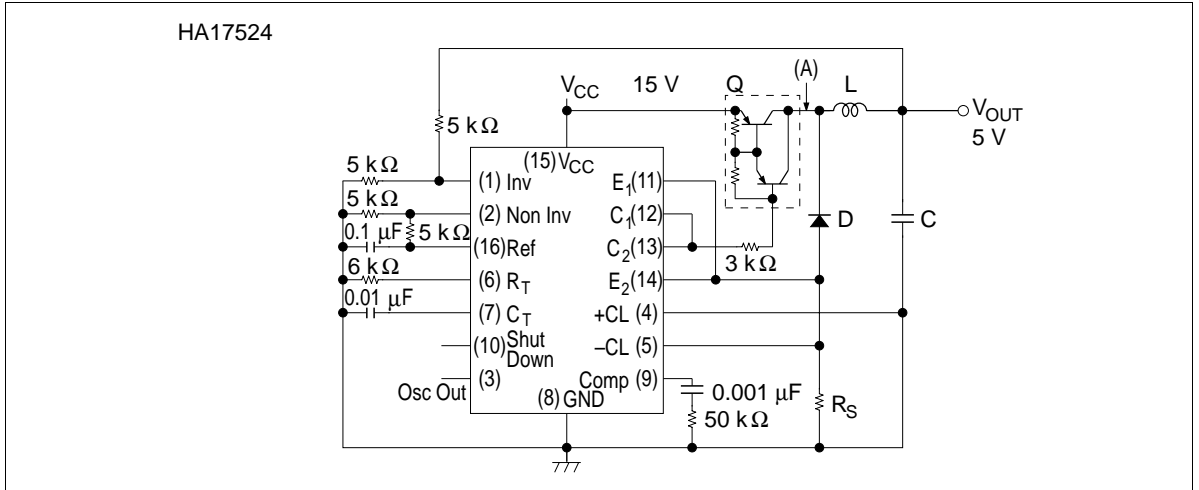


Figure 8 Stepdown Voltage Type Chopper Switching Regulator

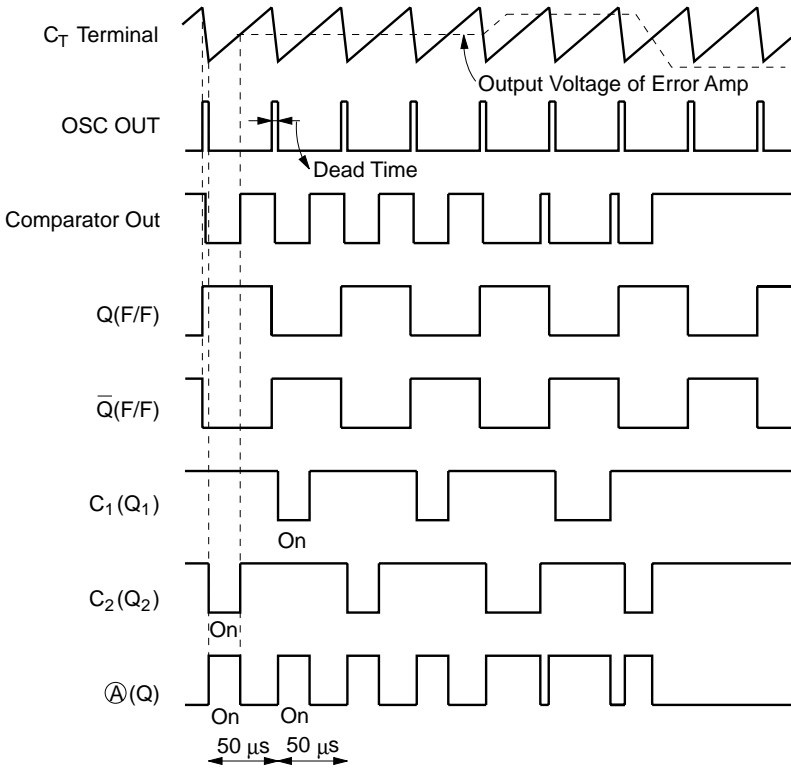


Figure 9 Operating Waveforms

### Circuit Applications

Simplified inverting Regulator: Figure 10 shows the circuit configuration of HA17524 inverting regulator for light load ( $V_{OUT} = -5\text{ V}$ )

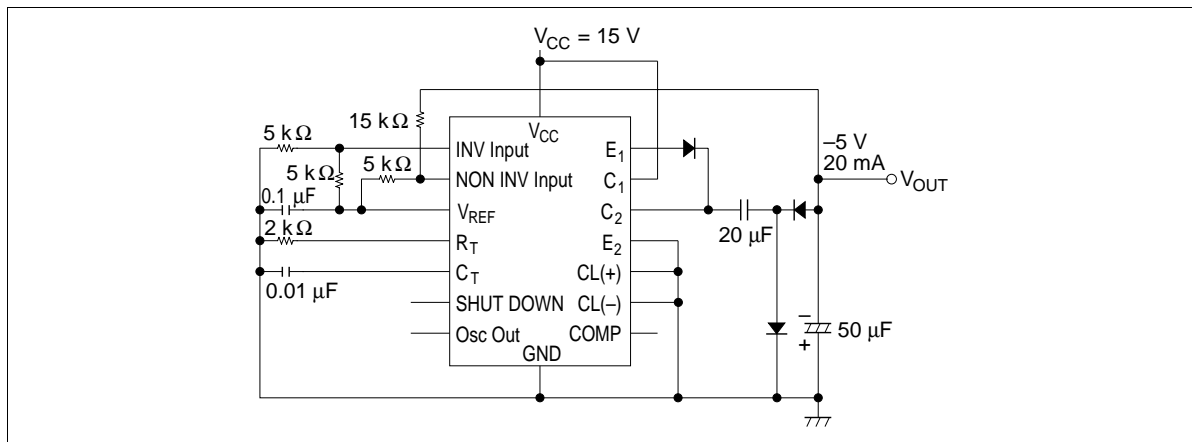


Figure 10 Simple Polarity Conversion

Tracking Switching Regulator: Figure 11 shows the circuit configuration of a tracking regulator that uses a transformer. ( $V_{OUT} = \pm 15\text{ V}$ )

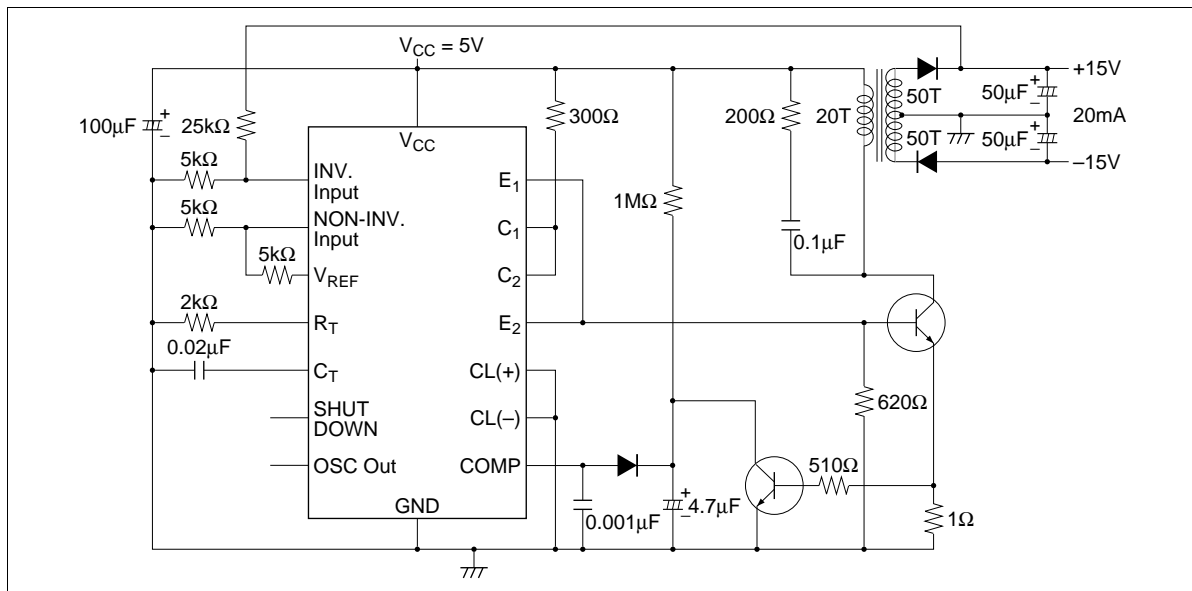
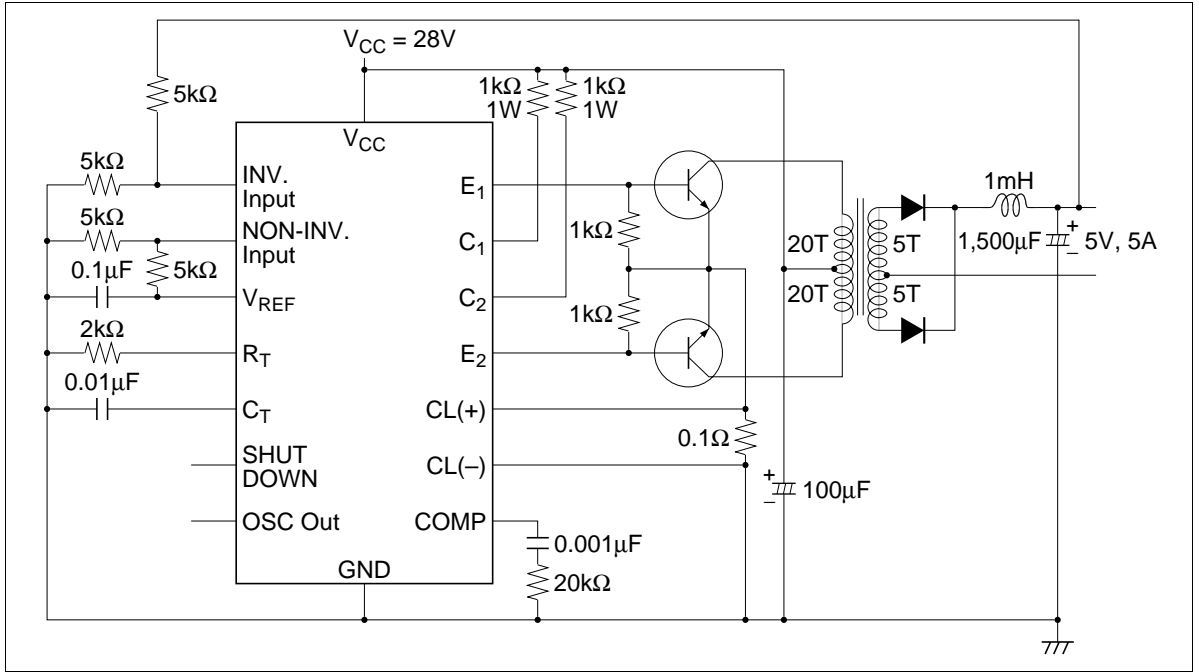


Figure 11 Tracking Switching Regulator

Push Pull Switching Regulator: Figure 12 shows the circuit configuration of push-pull switching regulator that uses transformer. This system is suited for high power. Output transistors inside HA17524 can drive external switching transistors.



**Figure 12 Push-Pull Switching Regulator**

## Note

Compared with conventional series regulators, switching regulators generate high frequency noise by switching current quickly. To reduce noise

1. As a general rule, insert line filter to reduce noise at the input.
2. To reduce noise at the output:
  - a. Twist output wiring together.
  - b. Do not bundle power source and output wiring.
  - c. Insert capacitor should be inserted at the load side.
  - d. Ground the power frame.
3. When choosing external parts (external switching transistor, diode, coil, etc) consider their capacitance and characteristics.

**Absolute Maximum Ratings** (Unless otherwise specified,  $T_a = +25^\circ\text{C}$ )

Item	Symbol	Rating	Unit	Note
Supply voltage	$V_{CC}$	40	V	1, 2
Collector output current	$I_C$	100	mA	
Reference output current	$I_{REF}$	50	mA	
Current through $C_T$ terminal	$I_{CT}$	5	mA	
Continuous total power dissipation	$P_T$	600	mW	3
Operating free-air temperature range	$T_{opr}$	-20 to +75	$^\circ\text{C}$	
Storage temperature range	$T_{stg}$	-55 to +125	$^\circ\text{C}$	

Notes: 1. With respect to network ground terminal

2. The reference voltage can be given by connecting the  $V_{CC}$  and 5 V reference output pins both to the supply voltage. In this configuration,  $V_{CC} = 6\text{ V}$  max.

3. HA17524P: Value at  $T_a \leq 52.7^\circ\text{C}$ , If  $T_a > 52.7^\circ\text{C}$ , derate by  $8.3\text{ mW}/^\circ\text{C}$

## Electrical Characteristics ( $V_{CC} = 20\text{ V}$ , $f = 20\text{ kHz}$ , $T_a = 25^\circ\text{C}$ )

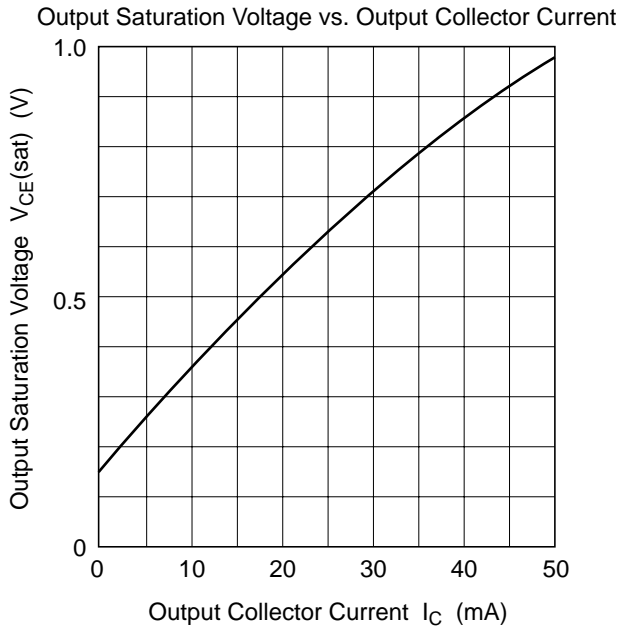
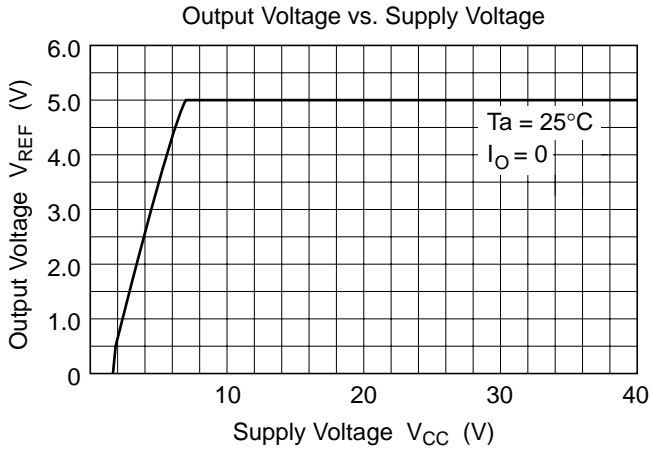
Item		Symbol	Min	Typ	Max	Unit	Test Conditions
Regulator	Output voltage	$V_{REF}$	4.6	5.0	5.4	V	
	Input regulation	$\delta V_{OLine}$	—	10	30	mV	$V_{CC} = 8\text{ to }40\text{ V}$
	Ripple rejection	$R_{REJ}$	—	66	—	dB	$f = 120\text{ Hz}$
	Output regulation	$\delta V_{OLoad}$	—	20	50	mV	$I_{out} = 0\text{ to }20\text{ mA}$
	Output voltage change with output temperature	$\delta V_O/\delta T_a$	—	0.3	1.0	%	$T_a = 0\text{ to }+70^\circ\text{C}$
			—	0.4	1.36	%	$T_a = -20\text{ to }+75^\circ\text{C}$
Short-circuit output current (Note)	$I_{OS}$	—	100	—	mA	$V_{REF} = 0$	
Error amplifier	Input offset voltage	$V_{IO}$	—	2	10	mV	$V_{IC} = 2.5\text{ V}$
	Input bias current	$I_I$	—	2	10	$\mu\text{A}$	$V_{IC} = 2.5\text{ V}$
	Open-loop voltage gain	$A_{VD}$	—	60	—	dB	
	Common-mode input voltage range	$V_{CM}$	1.8 to 3.4	—	—	V	$T_a = 25^\circ\text{C}$
	Common-mode Rejection ratio	CMR	—	70	—	dB	
	Unity-gain bandwidth	BW	—	3	—	MHz	
	Output swing	$V_{OPP}$	0.5	—	3.8	V	
Oscillator	OSC frequency	$f$	—	450	—	kHz	$C_T = 0.001\ \mu\text{F}$ , $R_T = 2\ \text{k}\Omega$
	Standard deviation of frequency	$\Delta f$	—	5	—	%	$V_{CC} = 8\text{ to }40\text{ V}$ , $R_T = 1.8\text{ to }100\ \text{k}\Omega$ , $C = \text{Const}$
	Frequency stability	$\delta f_{Line}$	—	—	1.0	%	$V_{CC} = 8\text{ to }40\text{ V}$
			—	5.0	10	%	$T_a = 0\text{ to }+70^\circ\text{C}$
			—	5.0	13.6	%	$T_a = -20\text{ to }+75^\circ\text{C}$
	Output amplitude	$V_{3(\text{peak})}$	—	3.5	—	V	Pin 3
Output pulse width	$T_p$	—	0.5	—	$\mu\text{s}$	$C_T = 0.01\ \mu\text{F}$ , Pin 3	
Comparator	Maximum duty cycle	Dmax	45	—	—	%	
	Threshold voltage	$V_{th\ 0}$	—	1.0	—	V	Duty cycle = 0
		$V_{th\ max}$	—	3.5	—	V	Duty cycle = max
	Input bias current	$I_I$	—	-1	—	$\mu\text{A}$	

Note: Duration of the short-circuit should not exceed one second.

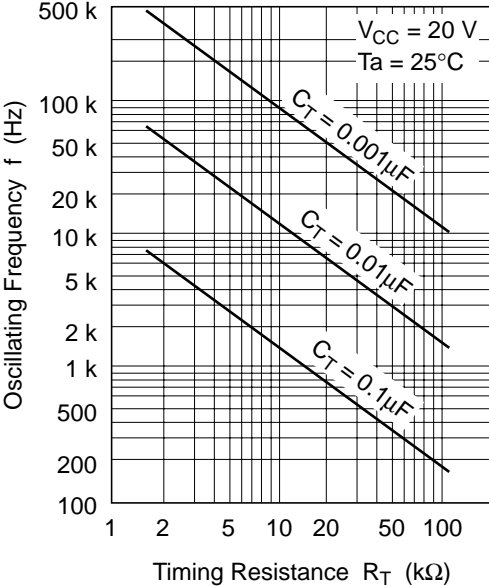
**Electrical Characteristics** ( $V_{CC} = 20\text{ V}$ ,  $f = 20\text{ kHz}$ ,  $T_a = 25^\circ\text{C}$ ) (cont)

Item		Symbol	Min	Typ	Max	Unit	Test Conditions
Current limiter	Input voltage range	$V_{IS}$	-0.7 to +1.0	—	—	V	
	Sense voltage	$V_S$	180	200	220	mV	$V(\text{Pin } 9) = 2\text{ V}$ , $T_a = 25^\circ\text{C}$ $V(\text{Pin } 2)$ $-V(\text{Pin } 1) \geq 50\text{ mV}$
	Sensevoltage change with temperature	$\delta V_S/\delta T_a$	—	0.2	—	mV/°C	$T_a = -20\text{ to }+75^\circ\text{C}$
Output	Collector-emitter breakdown voltage	$V_{CE}$	40	—	—	V	
	Collector off-state current	$I_{Leak}$	—	0.01	50	$\mu\text{A}$	$V_{CE} = 40\text{ V}$
	Collector-emitter saturation voltage	$V_{CE(sat)}$	—	1	2	V	$I_C = 50\text{ mA}$
	Emitter output voltage	$V_E$	17	18	—	V	$V_{CC} = 20\text{ V}$ , $I_E = -250\text{ }\mu\text{A}$
	Rise time	$t_r$	—	0.2	—	$\mu\text{s}$	$R_C = 2\text{ k}\Omega$
	Fall time	$t_f$	—	0.1	—	$\mu\text{s}$	
Total device	Standby current	$I_{ST}$	—	5.0	10	$\text{mA}$	$V_{CC} = 40\text{ V}$ , $V_2 = 2\text{ V}$ , Pins 1, 4, 7, 8, 9, 11, 14grounded, All other pins open

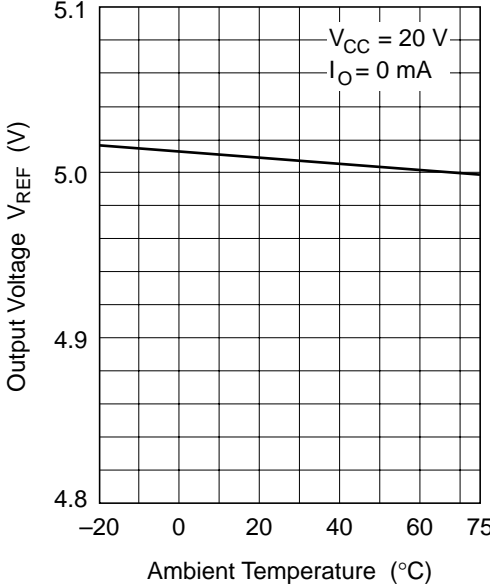
## Characteristic Curves

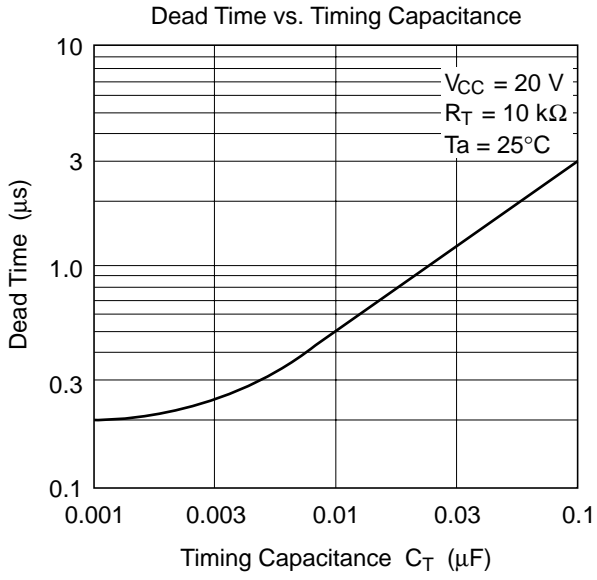


Oscillating Frequency vs. Timing Resistance



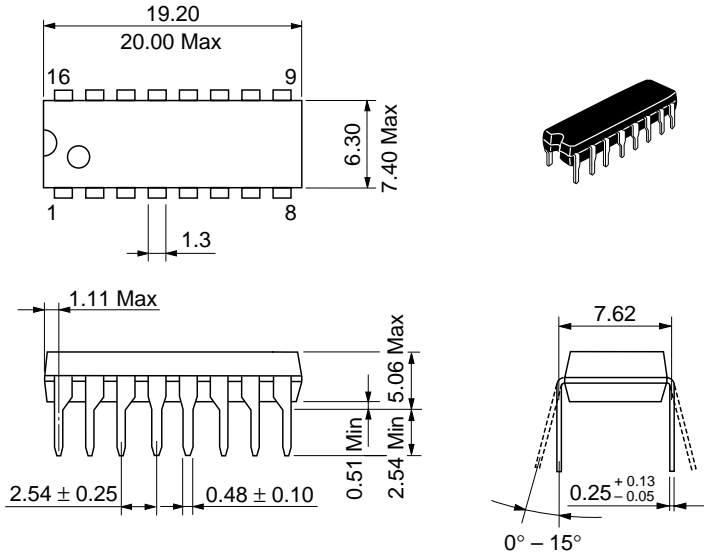
Output Voltage vs. Ambient Temperature





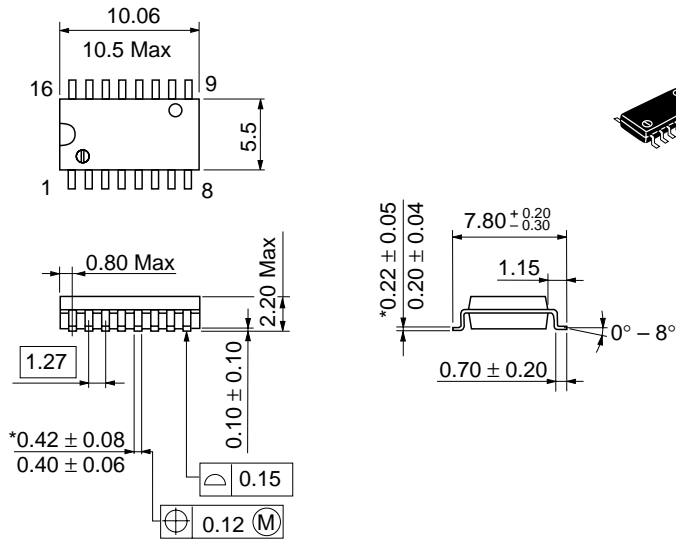
Package Dimensions

Unit: mm



Hitachi Code	DP-16
JEDEC	Conforms
EIAJ	Conforms
Mass (reference value)	1.07 g

Unit: mm



Hitachi Code	FP-16DA
JEDEC	—
EIAJ	Conforms
Mass (reference value)	0.24 g

\*Dimension including the plating thickness  
Base material dimension

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## Hitachi, Ltd.

Semiconductor & Integrated Circuits.  
Nippon Bldg., 2-6-2, Ohte-machi, Chiyoda-ku, Tokyo 100-0004, Japan  
Tel: Tokyo (03) 3270-2111 Fax: (03) 3270-5109

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### For further information write to:

Hitachi Semiconductor  
(America) Inc.  
179 East Tasman Drive,  
San Jose, CA 95134  
Tel: <1> (408) 433-1990  
Fax: <1> (408) 433-0223

Hitachi Europe GmbH  
Electronic Components Group  
Dornacher StraÙe 3  
D-85622 Feldkirchen, Munich  
Germany  
Tel: <49> (89) 9 9180-0  
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Hitachi Europe Ltd.  
Electronic Components Group.  
Whitebrook Park  
Lower Cookham Road  
Maidenhead  
Berkshire SL6 8YA, United Kingdom  
Tel: <44> (1628) 585000  
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Hitachi Tower  
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